$\underline{A^2\eta^2 n E_i(n)} \tau P_t \sigma f_{Tx}^2 \left| F^2 \right|^2$ $S/N = \frac{P_r}{P_{noise}} \frac{nE_i(n)B\tau}{L_{sp}}$ $4\pi c^2 L k_B T_o N_f R^4$

A bluffer's guide to Radar



Andy French December 2009



Nikola Tesla (1856-1943)

"We may produce at will, from a sending station, an electrical effect in any particular region of the globe; (with which) we may determine the relative position or course of a moving object, such as a vessel at sea, the distance traversed by the same, or its speed."



"And yes, my wig is very nice"

Nikola Tesla (1856-1943)

RAdio Detection And Ranging

Radars detect the presence of a physically remote object via the reception and processing of **backscattered electromagnetic** waves.

Unlike optical systems, (which are responsive to frequencies $\approx 10^{15}$ Hz), Radar is typically associated with frequency bands ranging from a **few MHz** (High Frequency or HF band) up to **hundreds of GHz** (mm wave).



 Most targets of interest (especially those constructed from metal) are highly reflective at Radar frequencies.

• Radar can be used in darkness and can penetrate haze, fog, snow and rain.

• Atmospheric propagation attenuation is much less severe for Radar than higher frequency electromagnetic disturbances. This means Radar can be used for long range surveillance. A military air defence system may have an operational range of hundreds of km.

• Radar has been used to successfully measure the distance between the Earth and other planets in the solar system. Note Mars is 56 million km from Earth!

I told you it would useful!

 The technology to generate, receive and process Radar signals has been continuously refined for nearly 100 years

 Military and civilian air traffic control have employed Radar as a key sensor extensively since the Second World War.





• Magnetron transmitters, which are stable sources of microwaves (0.1 - 100 GHz approximately) are ubiquitous as a fundamental element of modern domestic ovens.

• Given the size of a Radar antenna roughly scales with the wavelength it transmits / receives; Radars (with modest directivity, i.e. a beamwidth of a few degrees) tend to be of dimensions well suited to human use i.e. of the order of a few metres.







THE ELECTROMAGNETIC SPECTRUM







Radar bands

30-300Hz 3 - 30kHz 30 - 300kHz 300 - 3000kHz 30 - 300GHz 300GHz - 429THz 429 - 750THz >750THz

Extremely low frequency ELF Very low frequency VLF Low frequency LF **Medium frequency MF Extremely high frequency EHF** Infrared IR Visible Light **Ultraviolet UV**

Radar bands

The Radar Equation

First we state that girls require time and money.

And as we all know "time is money."

Time = Money Therefore:

 $(jirls = Money * Money = (Money)^2$ And because "money is the root of all evil":

> Money = JEVI Therefore:

Girls = (VEVI)2

And we are forced to conclude that:

Girls = Ev.





 $G \approx \frac{26,000}{(\Delta \epsilon/\mathrm{deg}) (\Delta \phi/\mathrm{deg})}$



 $G = \frac{4\pi A\eta}{\lambda^2}$ $\frac{1}{c^2} = \frac{4\pi A\eta}{c^2} f_{Tx}^2$

Antenna gain

 $G = \frac{4\pi A\eta}{\lambda^2} = \frac{4\pi A\eta}{c^2} f_{Tx}^2$

$G \approx \frac{26,000}{\left(\Delta \epsilon / \mathrm{deg}\right) \left(\Delta \phi / \mathrm{deg}\right)}$





$$P_{r} = \frac{P_{s}}{4\pi R^{2}} \frac{A\eta}{L_{r}} \left|F^{2}\right|^{2} = \frac{A^{2}\eta^{2}\sigma f_{Tx}^{2}}{4\pi R^{4}c^{2}} \frac{P_{t}}{L_{t}L_{r}} \left|F^{2}\right|^{2}$$

$$S/N = \frac{P_r}{P_{noise}} \frac{nE_i(n)B\tau}{L_{sp}} = \frac{A^2\eta^2 nE_i(n)\tau P_t \sigma f_{Tx}^2 \left|F^2\right|^2}{4\pi c^2 Lk_B T_o N_f R^4}$$

THE RADAR EQUATION

$$P_{noise} = k_B T_o B N_f$$

$$\begin{split} &A = 4\text{m}^2, \, \eta = 0.5, \, n = 8, \, E_i(8) = 1, \, \tau = 25.6 \mu\text{s}, \, P_t = 12 \text{kW}, \, \sigma = 20\text{m}^2, \, f_{Tx} = 3\text{GHz}, \, \left|F^2\right|^2 = 0\text{dB}, \\ &k_B = 1.38 \times 10^{-23}\text{JK}^{-1}, \, T_oN_f = 500\text{K}, \, R = 50\text{km}, \, L = 10\text{dB} \\ &\text{The Radar Equation above yields } S/N = 35.6\text{dB} \end{split}$$



Pulse Repetition Interval

Range processing & pulse compression

- Range profiles obtained by transmitting a frequency coded pulse and correlating received and transmitted signals
- Range resolution inversely proportional to pulse bandwidth B



Range samples & high range resolution



Stack of pulse compressor outputs for all frequency steps



Doppler shift

$$\phi = 2\pi f_{Tx} t_{delay}$$

$$t_{delay} = 2R/c \qquad \phi = \frac{4\pi f_{Tx} R}{c}$$

$$1 \quad d\phi \qquad 4\pi f \quad \dot{P}$$

$$f(t) = f_{Tx} - \frac{1}{2\pi} \frac{d\phi}{dt} \qquad \Delta\phi = \frac{4\pi f_{Tx}R}{cf_{PRF}}$$

$$f_D = -\frac{1}{2\pi} \frac{\Delta\phi}{1/f_{PRF}} = -\frac{2\dot{R}}{c} f_{Tx}$$

$$\begin{array}{l} \textbf{Doppler filter}\\ y = \sum_{p=1}^{P} \psi_p \overset{\text{Weights}}{\sum_{\text{Samples per pulse}}}\\ \psi_p = e^{2\pi i t_p f}\\ |y(f)|^2 = \left| \sum_{p=1}^{P} w_p e^{2\pi i t_p f} \right|^2 \overset{\text{GROUND}}{\sum_{\text{CLUTTER}}}\\ \{w_p\} = \{1, -2, 1\} \end{array}$$

Doppler filter: DFT

$$w_{p,k} = e^{-\frac{2\pi i (p-1)(k-1)}{P}}$$
Discrete fourier
TRANSFORM

$$y_k(f) = \left| \frac{e^{ix} \sin\left\{ \left(1 + \frac{1}{P}\right) x\right\}}{\sin\left(\frac{x}{P}\right)} - 1 \right|^2$$
$$\frac{x}{\pi} = 1 - k + \frac{fP}{f_{PRF}}$$

$$f_{\max} = \frac{k-1}{P} f_{PRF}$$

Doppler spectra

wfc2 Df=3p2 P=32 Q=32\DH8D_1023_BEE232_1048 AQ18 04.mat Doppler filter output. No JEM. Non skin energy = 47.8%

DOPPLER

FREQUENCY





Doppler spectrum for 32 pulse, 32 frequency step 2.5kHz PRF

Dash8 six blade propeller aircraft