



Multimeter (set in ACA mode i.e. gives RMS amplitude of (AC) current)



Capacitor block (non-electrolytic) range 0.220µF to 15µF.

2x green wires (PD across AC input, connected to oscilloscope CH1)

2x green wires (PD across capacitor, connected to oscilloscope CH2)

Fixed resistors (only 1000hm will be used)

Signal generator + mains cable kettle lead + BNC to 4mm converter

Dual input oscilloscope + mains cable kettle lead + 2 x BNC to 4mm converter

Capacitor block (nonelectrolytic) range 0.220µF to 15µF.







Fixed resistors (only 100ohm will be used)





Multimeter (set in ACA mode i.e. gives RMS amplitude of (AC) current)



4 x blue wires (main RC circuit loop)

2x green wires (PD across AC input, connected to oscilloscope CH1)

2x green wires (PD across capacitor, connected to oscilloscope CH2)



No external connection to ground, this is within the Sig Gen **AND** oscilloscope

Suggested experiments

- 1. Fix V_{in} at about **3.0V** (the middle value of the signal generator amplitude).
- 2. Fix *R* at about **100** Ω (measure it accurately with a multi-meter).
- 3. Fix capacitances at $1.0\mu F$, $2.2\mu F$ and $4.7\mu F$. Do the middle value only if time permits.
- 4. Record and plot gain $|V_c/V_{in}|$ vs frequency f for ten to twenty measurements over the frequency range **0** to **3,000Hz**.
- 5. Also record and plot the RMS current vs frequency.
- Fixing the frequency at 200Hz, find the RMS current for the *full range of capacitances*.
 You should find that current is proportional to capacitance.
- 7. For a 4.7µF capacitor, keep the frequency at 3,000Hz and switch to a square wave, and then a triangle wave output from the signal generator. Observe that the RC circuit *integrates* the input, if the output is taken across the capacitor. i.e. an output of triangle waves or parabolae, respectively.
- 8. See if you can observe a *differentiation* effect for low frequencies (i.e. about 100Hz) when the output is taken across the resistor instead of the capacitor.

Dual channel oscilloscope



Channel 1 (CH1): Potential difference across output of **signal generator** (upper trace). Channel 2 (CH2): Potential difference across **capacitor** (lower trace).





Characteristic frequency

 $f_c =$

Gain reduces as frequency increases.

Phase shift tends to 90° as frequency increases



When frequencies are much greater than the characteristic frequency, current tends to a *constant*:



Current is *proportional* to frequency (and capacitance) when frequencies are much less than the characteristic frequency.

When frequencies are much greater than the characteristic frequency, current tends to a *constant*:

RMS current vs capacitance. f=199.9Hz. $|V_{in}|$ =3V, R = 100 Ω .





Current is *proportional* to capacitance when frequencies are much less than the characteristic frequency.

Mathematical model of voltage gain vs frequency

 $\frac{V_{C}}{V_{in}} = \frac{Z_{C}}{Z_{C} + Z_{R}} \quad \text{Potential divider}$

$$\omega = 2\pi f \qquad Z_R = R$$
Complex impedence $Z_C = \frac{1}{i\omega C}$

$$\frac{V_{C}}{V_{in}} = \frac{1/i\omega C}{1/i\omega C + R} = \frac{1}{1 + i\omega RC} = \frac{1 - i\omega RC}{1 + \omega^{2}R^{2}C^{2}}$$

$$\begin{array}{c|c}
1 \\
\phi \\
\omega RC
\end{array}$$

$$\frac{V_{C}}{V_{in}} = \frac{1}{1 + \omega^{2} R^{2} C^{2}} \sqrt{1 + \omega^{2} R^{2} C^{2}} e^{-i \tan^{-1}(\omega RC)}$$
Argand (phasor) diagram
$$V_{C} = |V_{C}| e^{i(\omega t - \phi)} \quad V_{in} = |V_{in}| e^{i\omega t}$$
Argand (phasor) diagram
$$\frac{|V_{C}|}{|V_{in}|} = \frac{1}{\sqrt{1 + \omega^{2} R^{2} C^{2}}}$$

$$\phi = \tan^{-1}(\omega RC)$$

$$V_{C} = \frac{1}{\sqrt{1 + \omega^{2} R^{2} C^{2}}}$$

$$V_{in} = |V_{in}| e^{i\omega t}$$



Finding the RMS current

$$\omega = 2\pi f$$

$$V_C = \frac{\left|V_{in}\right|}{\sqrt{1 + \omega^2 R^2 C^2}} e^{i\left\{\omega t - \tan^{-1}(\omega RC)\right\}}$$

$$V_c = IZ_c$$

$$V_c = I/i\omega C$$
 : $I = i\omega CV_c$

$$\therefore I = \frac{i\omega C |V_{in}|}{\sqrt{1 + \omega^2 R^2 C^2}} e^{i\{\omega t - \tan^{-1}(\omega RC)\}}$$

$$\therefore |I| = \frac{|V_{in}|}{R} \times \frac{\omega RC}{\sqrt{1 + \omega^2 R^2 C^2}}$$

$$\therefore I_{RMS} = \frac{|V_{in}|}{\sqrt{2}R} \times \frac{\omega RC}{\sqrt{1 + \omega^2 R^2 C^2}}$$

Characteristic $f_c = \frac{1}{2\pi RC}$

Low frequency limit:

Current is proportional to frequency and capacitance.

$$I_{RMS} \approx \frac{\left|V_{in}\right|}{\sqrt{2R}} \omega RC \quad f \ll \frac{1}{2\pi RC}$$

High frequency limit:

Current tends to a constant, as one would expect for a DC circuit.





RC circuit as an *integrator*

High frequency limit

$$f \gg \frac{1}{2\pi RC}$$

$$V_{C} = \frac{\left|V_{in}\right|}{\sqrt{1 + \omega^{2}R^{2}C^{2}}} e^{i\left\{\omega t - \tan^{-1}(\omega RC)\right\}}$$

 $\omega RC \gg 1$

$$\Rightarrow V_{C} \approx \frac{\left|V_{in}\right|}{\omega RC} e^{i\{\omega t - \frac{\pi}{2}\}} = \frac{\left|V_{in}\right|}{i\omega RC} e^{i\omega t} = \frac{V_{in}}{i\omega RC} \qquad e^{i\frac{\pi}{2}} = i$$
$$e^{-i\frac{\pi}{2}} = 1/i$$
$$V_{in} = \left|V_{in}\right| e^{i\omega t}$$

$$\therefore \int V_{in} dt = |V_{in}| \int e^{i\omega t} dt = \frac{|V_{in}|}{i\omega} e^{i\omega t} + c$$
$$\therefore V_{C} \approx \frac{1}{RC} \int V_{in} dt$$

So a square wave input will produce a 'sawtooth' (triangular wave) capacitor voltage.



RC circuit as an *differentiator*

Low frequency limit

$$f \ll \frac{1}{2\pi RC}$$

Note: For an unknown reason
I couldn't get this to work, even
when using the
$$0.22\mu$$
F capacitor
and frequency turned down to 100Hz.
In this case $f_c = 6,786$ Hz.

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So a 'sawtooth' (triangular wave) input will result in a square wave resistor potential difference.

$$\frac{V_R}{V_{in}} = \frac{R}{1/i\omega C + R} = \frac{i\omega RC}{1 + i\omega RC}$$

$$\omega RC \ll 1$$

$$\Rightarrow V_R \approx i\omega RCV_{in}$$

$$V_{in} = |V_{in}| e^{i\omega t}$$

$$\therefore \frac{dV_{in}}{dt} = i\omega V_{in}$$

$$\therefore V_R = RC \frac{dV_{in}}{dt}$$

Excel sheet of measurements taken using RC circuit setup

	А	В	С	D	E	F	G	Н	I	J	K	L	М	N	0	Р	Q	R	S	Т	U	V	W	Х	Γ
1	Vin = 3.4V	R = 106.6 oh	ms																						Π
2																									
3																									
4	RC circuit exp	eriment. Dr A	French. Win	chester Co	lege la	boratory F	25. 26/09/2	020.																	
5												RMS	current	(I) vs fre	qency		Vin	/v	3.4						
6	For all data p	or all data points below:			R	₹ = 106.7 ohms							25.00							f/H	z	199.9			
7																									
8	C = 4.7muF			C = 3	2.2muF	uF		C = 1.0m		uF			20.00							C /m	nuF	I RMS /mA	1/(2*pi*R*C) /Hz		
9													₹	-						0.22		0.65	6786.4		
10	f /hz	I RMS /mA	Vc /volts	f/h	z I	RMS /mA	Vc /volts		f /hz	I RMS /mA	Vc /volts		۳.							0.47		1.42	3176.6		
11	43	3.03	3.30	42.4	14 1	.39	3.40		42.44	0.68	3.40			A /	~					1		3.15	1493.0		
12	102.7	6.60	3.25	99.1	L 3	3.12	3.30		99.1	1.54	3.35		rer		-					2.2		6.05	678.6		
13	150	9.38	3.00	151	.3 4	.76	3.25		151.3	2.34	3.40		ਤੋਂ 10.00							4.7		11.56	317.7		
14	200.2	11.84	2.80	202	.3 6	5.13	3.20		202.3	3.10	3.30		RM	I K						6.8		14.55	219.6		
15	264.2	14.29	2.60	265	.9 7	.85	3.10		265.8	4.01	3.30		5.00							10		17.27	149.3		
16	316.4	15.20	2.30	321	9	.22	3.02		321	4.87	3.30									15		18.64	99.5		
17	412.9	17.00	2.00	408	.5 1	1.10	2.90		408.5	6.13	3.20		0.00	1											
18	503.1	17.88	1.73	504	.7 1	2.86	2.70		504.7	7.30	3.17		(500	1000 1	500 2000) 2500	3000	3500						
19	678.6	19.15	1.40	681	.8 1	.4.99	2.40		681.8	9.36	3.00				fr	equency /H	z					I RMS v	s C (f=200Hz)		
20	761.5	19.30	1.23	768	.1 1	.5.49	2.20		768.1	10.17	2.93										20				
21	894.1	19.72	1.10	893	.9 1	.6.63	2.03		893.9	11.43	2.82			 4.	7muF 🔶	- 2,2muF —	●—1.0muF								
22	1058	21.10	1.00	105	6 1	.7.47	1.84		1056	12.74	2.70										18				
23	1155	21.50	0.93	114	6 1	.7.93	1.80		1146	13.24	2.60										16				
24	1285	21.50	0.85	128	4 1	.8.39	1.63		1284	13.95	2.40				Vc vs t	reqency					10				
25	1457	22.00	0.75	147	4 1	.8.63	1.40		1474	15.38	2.35		4.00								14	/	6		
26	1884	22.00	0.60	189	0 1	.9.09	1.12		1890	16.30	1.98		2.50												
27	2684	21.90	0.45	267	4 1	.9.79	0.83		2674	18.37	1.51		5.50	Con.						4	12 -	<u> </u>			
28	3006	21.80	0.40	302	5 2	0.70	0.80		3025	18.58	1.40		3.00		~					S /m	10	/			
29													2.50	- X	-					Z					
30													olts	- X S							8				
31													2.00		N .		<hr/>				6				
32													> 1.50								0				
33														-							4	/			
34													1.00				-					<i>†</i>			
35													0.50								2				
36													0.00					Ī			0				
37													0.00 0	500	1000 1	500 2000	2500	3000	3500		0	5	10 15	20	
38															fr	equency /H	z					C	apacitance /muF		
39																									
40													- 4.7muF - 2.2muF - 1.0muF - 1.0muF - 1.0muF												
41																									
42																									

As frequency and capacitance changes, *the amplitude of the signal generator also changes slightly*. In this example, the signal generator amplitude has to be changed for every measurement to ensure a consistent value of 3.4V. The latter was measured using the oscilloscope.





Experimental data underlaid with model curves









Frequency /Hz

C=4.7 μ F, f_c=0.282kHz

0,